Basic Antenna Theory

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Note: These are preliminary notes, intended only for distribution among the participants. Beware of misprints!
Purpose

• to refresh basic concepts related to the antenna physics
  – needed to understand better the operation and design of microwave links and systems
Outline

• Introduction
• Review of basic antenna types
• Radiation pattern, gain, polarization
• Equivalent circuit & radiation efficiency
• Smart antennas
• Some theory
• Summary
Quiz

We use a transmitting antenna to radiate radio waves and a receiving antenna to capture the RF energy carried by the waves.

Somebody told that the receiving antenna also radiates radio waves during the reception.

Is it a true fact or a slip of the tongue?
Intended & unintended radiators

- Antennas intended to produce specified EM field
  - Radiocommunication antennas; Measuring antennas; EM sensors, probes; EM applicators (Industrial, Medical, Scientific)
- Radiators not intended to generate any EM field, but producing it as an unintended side-effect
  - Any conductor/installation with varying electrical current (e.g. electrical installation of vehicles)
  - Any slot/opening in the screen of a device/cable carrying RF current
  - Any discontinuity in transmission medium (e.g. conducting structures/installations) irradiated by EM waves
    - Stationary (e.g. antenna masts or power line wires); Time-varying (e.g. windmill or helicopter propellers); Transient (e.g. aeroplanes, missiles)
Antenna purpose

- Transformation of a guided EM wave in transmission line (waveguide) into a freely propagating EM wave in space (or vice versa) with specified directional characteristics
  - Transformation from time-function in one-dimensional space into time-function in three dimensional space
  - The specific form of the radiated wave is defined by the antenna structure and the environment
Antenna functions

• Transmission line
  – Power transport medium - must avoid power reflections, otherwise use matching devices

• Radiator
  – Must radiate efficiently – must be of a size comparable with the half-wavelength

• Resonator
  – Unavoidable - for broadband applications resonances must be attenuated
Monopole (dipole over plane)

- If there is an inhomogeneity (obstacle) a reflected wave, standing wave, & higher field modes appear
- With pure standing wave the energy is stored and oscillates from entirely electric to entirely magnetic and back
- Model: a resonator with high $Q = \frac{\text{energy stored}}{\text{energy lost}}$ per cycle, as in LC circuits
- Kraus p.2
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Antennas for laptop applications

• Patch and slot antennas derived from printed-circuit and micro-strip technologies

• Ceramic chip antennas are typically helical or inverted-F (INF) antennas, or variations of these two types with high dielectric loading to reduce the antenna size

Slot & INF antennas

• Slot antenna: a slot is cut from a large (relative to the slot length) metal plate.
  • The center conductor of the feeding coaxial cable is connected to one side of the slot, and the outside conductor of the cable - to the other side of the slot.
  • The slot length is some ($\lambda/2$) for the slot antenna and ($\lambda/4$) long for the INF antenna.
• The slot and INF antennas behave similarly.
  • The slot antenna can be considered as a loaded version of the INF antenna. The load is a quarter-wavelength stub, i.e. a narrowband device.
  • When the feed point is moved to the short-circuited end of the slot (or INF) antenna, the impedance decreases. When it is moved to the slot center (or open end of the INF antenna), the impedance increases.
Example
do double-layer printed Yagi antenna

Note: no galvanic contact with the director

Source: N Gregorieva
• Patch and slot antennas are
  – Cheap and easy to fabricate and to mount
  – Suited for integration
  – Light and mechanically robust
  – Have low cross-polarization
  – Low-profile - widely used in antenna arrays
    – spacecrafts, satellites, missiles, cars and other mobile applications
Aperture-antenna

- Aperture antennas derived from waveguide technology (circular, rectangular)
- Can transfer high power (magnetrons, klystrons)
- Above few GHz
- Will be explored inprace during the school

Power absorbed: $P$ [watt]
Effective aperture: $A$ [$m^2$]

Power density: $PFD$ [$w/m^2$]

$A = A \times PFD$

Note: The aperture concept is applicable also to wired antennas. For instance, the max effective aperture of linear $\lambda/2$ wavelength dipole antenna is $\lambda^2/8$
Leaky-wave antennas

- Derived from millimeter-wave guides (dielectric guides, microstrip lines, coplanar and slot lines).
- For frequencies > 30 GHz, including infrared
- Subject of intensive study.
  - Note: Periodical discontinuities near the end of the guide lead to substantial radiation leakage (radiation from the dielectric surface).

Source: adapted from N Gregorieva

Property of R Struzak
Reflector antennas

• Reflectors are used to concentrate flux of EM energy radiated/ received, or to change its direction
• Usually, they are parabolic (paraboloidal).
  – The first parabolic (cylinder) reflector antenna was used by Heinrich Hertz in 1888.
• Large reflectors have high gain and directivity
  – Are not easy to fabricate
  – Are not mechanically robust
  – Typical applications: radio telescopes, satellite telecommunications.

Source: adapted from N Gregorieva
Planar reflectors

• Uda-Yagi, Log-periodic antennas

• Intended reflector antenna allows maintaining radio link in non-LOS conditions (avoiding propagation obstacles)

• Unintended antennas create interference
Paraboloidal reflectors

Front feed

Cassegrain feed
The largest radio telescopes

• Max Plank Institut für Radioastronomie radio telescope, Effelsberg (Germany), 100-m paraboloidal reflector
• The Green Bank Telescope (the National Radio Astronomy Observatory) – paraboloid of aperture 100 m
The Arecibo Observatory Antenna System

The world’s largest single radio telescope

304.8-m spherical reflector

National Astronomy and Ionosphere Center (USA), Arecibo, Puerto Rico
The Arecibo Radio Telescope

[Sky & Telescope
Feb 1997 p. 29]
Lens antennas

Lenses play a similar role to that of reflectors in reflector antennas: they collimate divergent energy. Often preferred to reflectors at frequencies > 100 GHz.

Source: Kraus p.382, N Gregorieva
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Radiation pattern

- The *radiation pattern of antenna* is a representation (pictorial or mathematical) of the distribution of the power out-flowing (radiated) from the antenna (in the case of transmitting antenna), or inflowing (received) to the antenna (in the case of receiving antenna) as a function of direction angles from the antenna

- Antenna radiation pattern (*antenna pattern*):
  - is defined for large distances from the antenna, where the spatial (angular) distribution of the radiated power does not depend on the distance from the radiation source
  - is independent on the power flow direction: it is the same when the antenna is used to transmit and when it is used to receive radio waves
  - is usually different for different frequencies and different polarizations of radio wave radiated/ received
Power pattern vs. Field pattern

- The *power pattern* is the measured (calculated) and plotted received power: $|P(\theta, \varphi)|$ at a constant (large) distance from the antenna.

- The *amplitude field pattern* is the measured (calculated) and plotted electric (magnetic) field intensity, $|E(\theta, \varphi)|$ or $|H(\theta, \varphi)|$ at a constant (large) distance from the antenna.

The power pattern and the field patterns are inter-related:

\[
P(\theta, \varphi) = \left(\frac{1}{\eta}\right)^2 |E(\theta, \varphi)|^2 = \eta^2 |H(\theta, \varphi)|^2
\]

- $P = $ power
- $E = $ electrical field component vector
- $H = $ magnetic field component vector
- $\eta = 377$ ohm (free-space, plane wave impedance)
Normalized pattern

• Usually, the pattern describes the normalized field (power) values with respect to the maximum value.
  – Note: The power pattern and the amplitude field pattern are the same when computed and when plotted in dB.
3-D pattern

- Antenna radiation pattern is 3-dimensional
- The 3-D plot of antenna pattern assumes both angles $\theta$ and $\varphi$ varying, which is difficult to produce and to interpret

Source: NK Nikolova
2-D pattern

- Usually the antenna pattern is presented as a 2-D plot, with only one of the direction angles, \( \theta \) or \( \varphi \) varies
- It is an intersection of the 3-D one with a given plane
  - usually it is a \( \theta = \text{const} \) plane or a \( \varphi = \text{const} \) plane that contains the pattern’s maximum

Source: NK Nikolova

Property of R Struzak
Example: a short dipole on z-axis

Elevation plane: $\varphi = \text{const}$

$\theta = 45^\circ$, $|\vec{r}| = 1/\sqrt{2}$

$|\vec{r}| = 1$

Source: NK Nikolova

Property of R Struzak
Principal patterns

- Principal patterns are the 2-D patterns of linearly polarized antennas, measured in 2 planes
  1. the \textit{E-plane}: a plane parallel to the $E$ vector and containing the direction of maximum radiation, and
  2. the \textit{H-plane}: a plane parallel to the $H$ vector, orthogonal to the $E$-plane, and containing the direction of maximum radiation

Source: NK Nikolova

Property of R Struzak
Example

Source: NK Nikolova

Property of R Struzak
Isotropic antenna

- **Isotropic antenna or isotropic radiator** is a hypothetical (not physically realizable) concept, used as a useful reference to describe real antennas.
- Isotropic antenna radiates equally in all directions.
  - Its radiation pattern is represented by a sphere whose center coincides with the location of the isotropic radiator.

Source: NK Nikolova
Directional antenna

• *Directional antenna* is an antenna, which radiates (or receives) much more power in (or from) some directions than in (or from) others.
  – Note: Usually, this term is applied to antennas whose directivity is much higher than that of a half-wavelength dipole.

Source: NK Nikolova
Omnidirectional antenna

- An antenna, which has a non-directional pattern in a plane
  - It is usually directional in other planes

Source: NK Nikolova

Property of R Struzak
Pattern lobes

*Pattern lobe* is a portion of the radiation pattern with a local maximum. Lobes are classified as: major, minor, side lobes, back lobes.

Source: NK Nikolova
Pattern lobes and beam widths

Source: NK Nikolova

Property of R Struzak
Example

Source: NK Nikolova
Beamwidth

• *Half-power beamwidth* (HPBW) is the angle between two vectors from the pattern’s origin to the points of the major lobe where the radiation intensity is half its maximum
  • Often used to describe the antenna resolution properties
    » Important in radar technology, radioastronomy, etc.

• *First-null beamwidth* (FNBW) is the angle between two vectors, originating at the pattern’s origin and tangent to the main beam at its base.
  » Often FNBW $\approx 2 \times$ HPBW
Antenna Mask (Example 1)

Typical relative directivity-mask of receiving antenna (Yagi ant., TV dcm waves)

[CCIR doc. 11/645, 17-Oct 1989]
Antenna Mask (Example 2)

Reference pattern for co-polar and cross-polar components for satellite transmitting antennas in Regions 1 and 3 (Broadcasting ~12 GHz)
Anisotropic sources: gain

- Every real antenna radiates more energy in some directions than in others (i.e. has directional properties).
- Idealized example of directional antenna: the radiated energy is concentrated in the yellow region (cone).
- Directive antenna gain: the power flux density is increased by (roughly) the inverse ratio of the yellow area and the total surface of the isotropic sphere.
  - Gain in the field intensity may also be considered - it is equal to the square root of the power gain.
Antenna gain measurement

Step 1: reference

\[ P_o = \text{Power delivered to the reference antenna} \]
\[ S_0 = \text{Power received (the same in both steps)} \]

Step 2: substitution

\[ P = \text{Power delivered to the actual antenna} \]
\[ S = \text{Power received (the same in both steps)} \]

Antenna Gain = \((P/P_o) \quad S=S_0\)
Antenna Gains $G_i, G_d$

- Unless otherwise specified, the gain refers to the direction of maximum radiation.
- Gain is a dimension-less factor related to power and usually expressed in decibels.
- $G_i$ “Isotropic Power Gain” – theoretical concept, the reference antenna is isotropic.
- $G_d$ - the reference antenna is a half-wave dipole.
## Typical Gain and Beamwidth

<table>
<thead>
<tr>
<th>Type of antenna</th>
<th>$G_i \text{ [dB]}$</th>
<th>BeamW.</th>
</tr>
</thead>
<tbody>
<tr>
<td>Isotropic</td>
<td>0</td>
<td>$360^0 \times 360^0$</td>
</tr>
<tr>
<td>Half-wave Dipole</td>
<td>2</td>
<td>$360^0 \times 120^0$</td>
</tr>
<tr>
<td>Helix (10 turn)</td>
<td>14</td>
<td>$35^0 \times 35^0$</td>
</tr>
<tr>
<td>Small dish</td>
<td>16</td>
<td>$30^0 \times 30^0$</td>
</tr>
<tr>
<td>Large dish</td>
<td>45</td>
<td>$1^0 \times 1^0$</td>
</tr>
</tbody>
</table>
Gain, Directivity, Radiation Efficiency

- The radiation intensity, directivity and gain are measures of the ability of an antenna to concentrate power in a particular direction.
- Directivity relates to the power \textit{radiated} by antenna \( (P_0) \)
- Gain relates to the power \textit{delivered} to antenna \( (P_T) \)

\[
G(\theta, \varphi) = \eta D(\theta, \varphi)
\]

\[
\eta = \frac{P_T}{P_0}
\]

- \( \eta \): radiation efficiency (0.5 - 0.75)
Antenna gain and effective area

• Measure of the effective absorption area presented by an antenna to an incident plane wave.

• Depends on the antenna gain and wavelength

\[ A_e = \eta \frac{\lambda^2}{4\pi} G(\theta, \varphi) \quad [m^2] \]

Aperture efficiency: \( \eta_a = \frac{A_e}{A} \)

A: physical area of antenna’s aperture, square meters
Power Transfer in Free Space

\[ P_R = PFD \cdot A_e \]
\[ = \left( \frac{G_T P_T}{4\pi r^2} \right) \left( \frac{\lambda^2 G_R}{4\pi} \right) \]
\[ = P_T G_T G_R \left( \frac{\lambda}{4\pi r} \right)^2 \]

- \( \lambda \): wavelength [m]
- \( P_R \): power available at the receiving antenna
- \( P_T \): power delivered to the transmitting antenna
- \( G_R \): gain of the transmitting antenna in the direction of the receiving antenna
- \( G_T \): gain of the receiving antenna in the direction of the transmitting antenna
- Matched polarizations
e.i.r.p.

• Equivalent Isotropically Radiated Power (in a given direction):

\[ e.i.r.p. = PG_i \]

• The product of the power supplied to the antenna and the antenna gain (relative to an isotropic antenna) in a given direction
Linear Polarization

• In a linearly polarized plane wave the direction of the E (or H) vector is constant.

• [http://www.amanogawa.com/archive/wavesA.html](http://www.amanogawa.com/archive/wavesA.html)
Elliptical Polarization

\[ Ex = \cos (wt) \]
\[ Ey = \cos (wt + \pi/4) \]

\[ Ex = \cos (wt) \]
\[ Ey = -\sin (wt) \]

\[ Ex = \cos (wt) \]
\[ Ey = \cos (wt + 3\pi/4) \]

\[ Ex = \cos (wt) \]
\[ Ey = -\cos (wt + \pi/4) \]

\[ Ex = \cos (wt) \]
\[ Ey = \sin (wt) \]
Polarization ellipse

- The superposition of two plane-wave components results in an elliptically polarized wave.
- The polarization ellipse is defined by its axial ratio $N/M$ (ellipticity), tilt angle $\psi$ and sense of rotation.
Polarization states

UPPER HEMISPHERE: ELLIPTIC POLARIZATION
LEFT_HANDED SENSE

EQUATOR: LINEAR POLARIZATION

LOWER HEMISPHERE: ELLIPTIC POLARIZATION
RIGHT_HANDED SENSE

POLES REPRESENT CIRCULAR POLARIZATIONS

LHC

LATITUDE: REPRESENTS AXIAL RATIO

45° LINEAR

LONGITUDE: REPRESENTS TILT ANGLE

(Poincaré sphere)

Property of R Struzak
Comments on Polarization

• At any moment in a chosen reference point in space, there is actually a single electric vector E (and associated magnetic vector H).
• This is the result of superposition (addition) of the instantaneous fields E (and H) produced by all radiation sources active at the moment.
• The separation of fields by their wavelength, polarization, or direction is the result of ‘filtration’.
Antenna Polarization

• The polarization of an antenna in a specific direction is defined to be the polarization of the wave produced by the antenna at a great distance at this direction
Polarization Efficiency

• The power received by an antenna from a particular direction is maximal if the polarization of the incident wave and the polarization of the antenna in the wave arrival direction have:
  – the same axial ratio
  – the same sense of polarization
  – the same spatial orientation
Polarization filters/ reflectors

Wall of thin parallel wires (conductors)

|E₁| > 0  
|E₂| = 0  
Vector E || wires

|E₁| > 0  
|E₂| ~ |E₂|  
Vector E ⊥ wires

Wire distance ~ 0.1λ

- At the surface of ideal conductor the tangential electrical field component = 0

Property of R Struzak
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The transmitter with the transmission line is represented by an (Thevenin) equivalent generator.

The antenna is represented by its input impedance (which is frequency-dependent and is influenced by objects nearby) as seen from the generator.

\[ jX_A \] represents energy stored in electric (\( E_e \)) and magnetic (\( E_m \)) near-field components; if \( |E_e| = |E_m| \) then \( X_A = 0 \) (antenna resonance).

\[ R_r \] represents energy radiated into space (far-field components).

\[ R_l \] represents energy lost, i.e. transformed into heat in the antenna structure.
Power transfer: Tx antenna

Transmitter is represented by an equivalent generator with \( V_G, R_G, X_G = const. \)

Let \( R_A = R_R + R_L; \quad R_A, X_A = \text{var.} \)

The power absorbed by antenna \( P = I^2 R_A \)

\[
I^2 = \frac{V_G}{\left[ \sqrt{(R_G + R_A)^2 + (X_G + X_A)^2} \right]^2}
\]

\[
P = V_G^2 \frac{R_A}{(R_G + R_A)^2 + (X_G + X_A)^2}
\]

\[
P = \left( \frac{V_G^2}{R_G} \right) \frac{R_A}{\frac{R_A}{R_G}} \left( 1 + \frac{R_A}{R_G} \right)^2 \left( \frac{X_G}{R_G} + \frac{X_A}{R_G} \right)^2
\]
\[ P = V_G^2 \frac{R_A}{(R_G + R_A)^2 + X_G^2 + 2X_G X_A + X_A^2} \]

\[ \frac{\partial P}{\partial X_A} = V_G^2 \left( -\frac{R_A(2X_G + 2X_A)}{\left[(R_G + R_A)^2 + (X_G + X_A)^2\right]^2} \right) \]

\[ \frac{\partial P}{\partial X_A} = 0, \text{ when } X_A = -X_G \]

*Maximum*: \[ \frac{\partial P}{\partial R_A} + \frac{\partial P}{\partial X_A} = 0 \]

\[ R_A = R_G, \quad X_A = -X_G \]

\[ P = \frac{V_G^2}{4R_G} \]

Let \( X_G + X_A = 0 \). Then

\[ P = V_G^2 \frac{R_A}{(R_G + R_A)^2} \]

\[ \frac{\partial P}{\partial R_A} = V_G^2 \left( \frac{(R_G + R_A)^2 - R_A^2(R_G + R_A)}{\left[(R_G + R_A)^2\right]^2} \right) = V_G^2 \left( \frac{R_G^2 + 2R_G R_A + R_A^2 - 2R_G R_A - 2R_A^2}{\left[(R_G + R_A)^2\right]^2} \right) \]

\[ \frac{\partial P}{\partial R_A} = 0, \text{ when } R_G = R_A \]
Impedance matching

\[ R_A = R_r + R_l = R_g \]
\[ X_A = -X_g \]
\[ P_A = \frac{|V_g|^2}{4R_A} \]
\[ P_g = \frac{|V_g|^2}{4R_g} \quad (= P_A) \]
\[ P_r = P_A \frac{R_r}{(R_r + R_l)} \]
\[ P_l = P_A \frac{R_l}{(R_r + R_l)} \]
Power vs. field strength

\[ P_r = \frac{|E|^2}{Z_0} \rightarrow |E| = \sqrt{P_r Z_0} \]

\[ |E| = \sqrt{E_\theta^2 + E_\phi^2} \]

\[ |H| = \frac{|E|}{Z_0} \]

\[ Z_0 = 377 \text{ ohms} \]

for plane wave

in free space
Receiving antenna equivalent circuit

The antenna with the transmission line is represented by an (Thevenin) equivalent generator

The receiver is represented by its input impedance as seen from the antenna terminals (i.e. transformed by the transmission line)

$V_A$ is the (induced by the incident wave) voltage at the antenna terminals determined when the antenna is open circuited

Note: The antenna impedance is the same when the antenna is used to radiate and when it is used to receive energy

Property of R Struzak
Power transfer

- The maximum power is delivered to (or from) the antenna when the antenna impedance and the impedance of the equivalent generator (or load) are matched.

![Graph showing power transfer vs. impedance ratio]

Property of R Struzak
• When the impedances are matched
  – Half of the source power is delivered to the load and half is dissipated within the (equivalent) generator as heat
  – In the case of receiving antenna, a part \( (P_j) \) of the power captured is lost as heat in the antenna elements, the other part being reradiated (scattered) back into space
  • Even when the antenna losses tend to zero, still only half of the power captured is delivered to the load (in the case of conjugate matching), the other half being scattered back into space
• When the antenna impedance is not matched to the transmitter output impedance (or to the receiver input impedance) or to the transmission line between them, impedance-matching devices must be used for maximum power transfer
• Inexpensive impedance-matching devices are usually narrow-band
• Transmission lines often have significant losses
Radiation efficiency

- The radiation efficiency $e$ indicates how efficiently the antenna uses the RF power.
- It is the ratio of the power radiated by the antenna and the total power delivered to the antenna terminals (in transmitting mode). In terms of equivalent circuit parameters:

$$e = \frac{R_r}{R_r + R_l}$$
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Antenna arrays

- Consist of multiple (usually identical) antennas (elements) ‘collaborating’ to synthesize radiation characteristics not available with a single antenna. They are able
  - to match the radiation pattern to the desired coverage area
  - to change the radiation pattern electronically (electronic scanning) through the control of the phase and the amplitude of the signal fed to each element
  - to adapt to changing signal conditions
  - to increase transmission capacity by better use of the radio resources and reducing interference

- Complex & costly
  - Intensive research related to military, space, etc. activities
    » Smart antennas, signal-processing antennas, tracking antennas, phased arrays, etc.

Source: adapted from N Gregorieva
Satellite antennas (TV)

- Not an array!
The Earth’s atmosphere is transparent in the narrow visible-light window (4000-7000 angstroms) and the radio band between 1 mm and 10 m.

[Sky & Telescope Feb 1997 p.26]
The New Mexico Very Large Array

27 antennas along 3 railroad tracks provide baselines up to 35 km. Radio images are formed by correlating the signals garnered by each antenna.
2 GHz adaptive antenna

- A set of 48 2GHz antennas
  - Source: Arraycomm
Phased Arrays

• Array of N antennas in a linear or two-dimensional configuration + beam-forming & control device

• The amplitude and phase excitation of each individual antenna controlled electronically (“software-defined”)
  – Diode phase shifters
  – Ferrite phase shifters

• Inertia-less beam-forming and scanning (μsec) with fixed physical structure
• **Switched beam antennas**
  – Based on switching function between separate directive antennas or predefined beams of an array

• **Space Division Multiple Access (SDMA)** = allocating an angle direction sector to each user
  – In a TDMA system, two users will be allocated to the same time slot and the same carrier frequency
  – They will be differentiated by different direction angles
• **Dynamically phased array (PA):**
  – A generalization of the switched lobe concept
  – The radiation pattern continuously track the designated signal (user)
  – Include a *direction of arrival* (DoA) tracking algorithm
Beam Steering

- Beam-steering using phase shifters at each radiating element

Equi-phase wave front

\[ \Delta = \left( \frac{2\pi}{\lambda} \right) d \sin \theta \]

Radiating elements

Phase shifters

Power distribution
4-Bit Phase-Shifter (Example)

Alternative solution: Transmission line with controlled delay
Switched-Line Phase Bit

Phase bit = delay difference

Property of R Struzak
Simulation

- **2 omnidirectional antennas** (equal amplitudes)
  - Variables
    - distance increment
    - phase increment

- **N omnidirectional antennas**
  - Group factor (N omnidirectional antennas uniformly distributed along a straight line, equal amplitudes, equal phase increment)

- [http://www.amanogawa.com/archive/TwoDipole/Antenna2-2.html](http://www.amanogawa.com/archive/TwoDipole/Antenna2-2.html) (more details)
2 omnidirectional antennas

\[ D = 0.5\lambda, \theta = 0^0 \]

\[ D = 0.5\lambda, \theta = 90^0 \]

\[ D = 0.5\lambda, \theta = 180^0 \]
N omnidirectional antennas

- Array gain (line, uniform, identical power)
Antenna Arrays: Benefits

• Possibilities to control electronically
  – Direction of maximum radiation
  – Directions (positions) of nulls
  – Beam-width
  – Directivity
  – Levels of sidelobes
using standard antennas (or antenna collections) independently of their radiation patterns

• Antenna elements can be distributed along straight lines, arcs, squares, circles, etc.
Adaptive ("Intelligent") Antennas

- Array of N antennas in a linear, circular, or planar configuration
- Used for selection signals from desired sources and suppress incident signals from undesired sources
- The antenna pattern track the sources
- It is then adjusted to null out the interferers and to maximize the signal to interference ratio (SIR)
- Able to receive and combine constructively multipath signals
• The amplitude/phase excitation of each antenna controlled electronically ("software-defined")

• The weight-determining algorithm uses a-priori and/or measured information to adapt antenna to changing environment

• The weight and summing circuits can operate at the RF and/or at an intermediate frequency
Antenna sitting

• Radio horizon
• Effects of obstacles & structures nearby
• Safety
  – operating procedures
  – Grounding
    • lightning strikes
    • static charges
  – Surge protection
    • lightning searches for a second path to ground
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Maxwell’s Equations

• EM field interacting with the matter

• Reciprocity Theorem
  – Antenna characteristics do not depend on the direction of energy flow. The impedance & radiation pattern are the same when the antenna radiates signal and when it receives it.
  – Note: This theorem is valid only for linear passive antennas (i.e. antennas that do not contain nonlinear and unilateral elements, e.g. amplifiers)
EM Field of Current Element

\[ \vec{E} = \vec{E}_r + \vec{E}_\varrho + \vec{E}_\varphi \]

\[ \vec{H} = \vec{H}_r + \vec{H}_\varrho + \vec{H}_\varphi \]

\[ |E| = \sqrt{|E_r|^2 + |E_\varrho|^2 + |E_\varphi|^2} \]

\[ |H| = \sqrt{|H_r|^2 + |H_\varrho|^2 + |H_\varphi|^2} \]

*I*: current (monochromatic) [A]; *dz*: antenna element (short) [m]
Short dipole antenna: summary

- $E_\theta$ & $H_\theta$ are maximal in the equatorial plane, zero along the antenna axis
- $E_r$ is maximal along the antenna axis $dz$, zero in the equatorial plane
- All show axial symmetry
- All are proportional to the current moment $ldz$
- Have 3 components that decrease with the distance-to-wavelength ratio as
  - $(r/\lambda)^{-2}$ & $(r/\lambda)^{-3}$: near-field, or induction field. The energy oscillates from entirely electric to entirely magnetic and back, twice per cycle. Modeled as a resonant LC circuit or resonator;
  - $(r/\lambda)^{-1}$: far-field or radiation field
  - These 3 component are all equal at $(r/\lambda) = 1/(2\pi)$
Field components

C, Q: Induction fields

FF: Radiation field

Relative field strength vs. Relative distance, $Br$
Field impedance

Field impedance
$Z = \frac{E}{H}$
depends on the antenna type and on distance.

Property of R Struzak
Far-Field, Near-Field

- Near-field region:
  - Angular distribution of energy depends on distance from the antenna;
  - Reactive field components dominate (L, C)

- Far-field region:
  - Angular distribution of energy is independent on distance;
  - Radiating field component dominates (R)
  - The resultant EM field can locally be treated as uniform (TEM)
Poynting vector

• The time-rate of EM energy flow per unit area in free space is the *Poynting vector* (see [http://www.amanogawa.com/archive/docs/EM8.pdf](http://www.amanogawa.com/archive/docs/EM8.pdf)).

• It is the cross-product (vector product, right-hand screw direction) of the electric field vector (E) and the magnetic field vector (H): \( \mathbf{P} = \mathbf{E} \times \mathbf{H} \).

• For the elementary dipole \( E_\theta \perp H_\theta \) and only \( E_\theta \times H_\theta \) carry energy into space with the speed of light.
Power Flow

• In free space and at large distances, the radiated energy streams from the antenna in radial lines, i.e. the Poynting vector has only the radial component in spherical coordinates.

• A source that radiates uniformly in all directions is an *isotropic source (radiator, antenna)*. For such a source the radial component of the Poynting vector is independent of $\theta$ and $\phi$. 
Linear Antennas

- Summation of all vector components $E$ (or $H$) produced by each antenna element:
  
  $E = E_1 + E_2 + E_3 + \ldots$

  $H = H_1 + H_2 + H_3 + \ldots$

- In the far-field region, the vector components are parallel to each other.

- Phase difference due to:
  - Excitation phase difference
  - Path distance difference

- Method of moments
Simulation: Linear dipole antenna

  - Linear dipole antenna
- [http://www.amanogawa.com/archive/Antenna1/Antenna1-2.html](http://www.amanogawa.com/archive/Antenna1/Antenna1-2.html)
  - Detailed analysis
Point Source

• For many purposes, it is sufficient to know the direction (angle) variation of the power radiated by antenna at large distances.
• For that purpose, any practical antenna, regardless of its size and complexity, can be represented as a point-source.
• The actual field near the antenna is then disregarded.
• The EM field at large distances from an antenna can be treated as originated at a point source - fictitious volume-less emitter.

• The EM field in a homogenous unlimited medium at large distances from an antenna can be approximated by an uniform plane TEM wave
Image Theory

- Antenna above perfectly conducting plane surface
- Tangential electrical field component $= 0$
  - vertical components: the same direction
  - horizontal components: opposite directions
- The field (above the ground) is the same as if the ground is replaced by an mirror image of the antenna
- [http://www.amanogawa.com/archive/wavesA.html](http://www.amanogawa.com/archive/wavesA.html)

Elliptical polarization: change of the rotation sense!
Summary

• Introduction
• Review of basic antenna types
• Radiation pattern, gain, polarization
• Equivalent circuit & radiation efficiency
• Smart antennas
• Some theory
Selected References

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Any questions?

Thank you for your attention
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